

#### **Features**

Internal Precision Voltage Reference

◆ Accuracy: 0.2% (max)

◆ Temp Drift: 50ppm (max)

♦ Pin Selectable for 2.5V, 1.65V, and Ratio Metric

Switching Current HALL Sensor Excitation

◆ Reduce HALL Sensor Offset and Drift

♦ Reduce HALL Sensor 1/f Noise

Extended Current Measurement Range

H-Bridge Drive Capability: 350 mA

Precision Current Sensing Amplifier

Offset and Drift: 100μV and 2μV/°C (Max)

◆ Bandwidth: 200kHz

Overrange and Error Flags

Power Supply: 2.7V to 5.5V

Package: QFN 4mm x 4mm with PowerPAD

Temp Range: -40°C to +125°C

# **Applications**

Close-Loop HALL Sensor Module

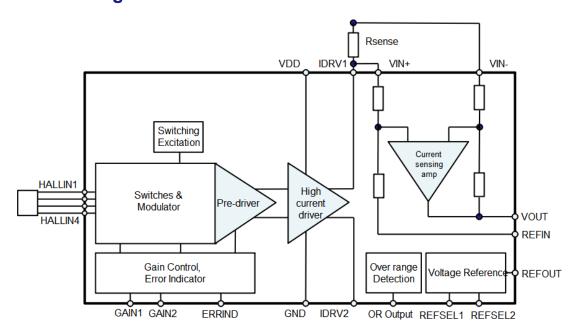
## **Description**

The TPAFE411 is designed only for Hall elements to use in closed-loop current-sensor modules. The internal precision excitation circuitry for the Hall-element could effectively eliminate the offset and offset-drift of the Hall element. The TPAFE411 also provides a 350mA H-bridge for driving the sensor compensation coil, and an internal precision current sensing amplifier to generate the output signal.

There is a high accuracy voltage reference, high accuracy HALL Sensor front end and precision current sensing amplifier inside TPAFE411. These techniques significantly improve the accuracy of the overall current-sensor module. When the power supply is 5V, the output voltage is pin-selectable to support a 2.5V output. When the power supply is 3.3V, the output voltage is to support a 1.65V output.

For the heat dissipation, 4mm × 4-mm QFN package with PowerPAD is selected for the TPAFE411. The TPAFE411 is specified to work over industrial temperature range of –40°C to +125°C.

# **Function Block Diagram**





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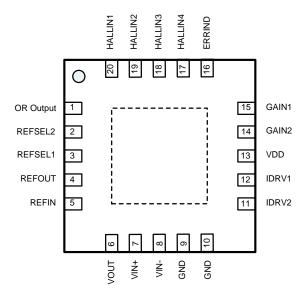


# **Revision History**

Date	Revision	Notes
2020/4/25	Rev.Pre.0	Pre-Release Version
2021/5/16	Rev.Pre.1	Pre-Release Version 1
2021/6/2	Rev.A.0	Release Version
2021/12/1	Rev.A.1	Update Electrical Characteristics table, fix typos



# **Pin Configuration and Pin Functions**



## **Pin Functions**

PIN Number	PIN Name	Description
1	OR Output	Open-drain output for overrange indication
2	REFSEL2	Reference mode select
3	REFSEL1	Reference mode select
4	REFOUT	Output for selected reference voltage
5	REFIN	Input for reference of current sensing amplifier
6	VOUT	Output of current sensing amplifier
7	VIN+	Noninverting input of current sensing amplifier
8	VIN-	Inverting input of current sensing amplifier
9	GND	Ground
10	GND	Ground
11	IDRV2	Output 2 of compensation coil driver
12	IDRV1	Output 1 of compensation coil driver
13	VDD	Power supply
14	GAIN2	Gain selected for HALL amplifier
15	GAIN1	Gain selected for HALL amplifier
16	ERRIND	Open-drain output for error indication
17	HALLIN4	Pin 4 of HALL sensor
18	HALLIN3	Pin 3 of HALL sensor
19	HALLIN2	Pin 2 of HALL sensor
20	HALLIN1	Pin 1 of HALL sensor
Thermal PAD		Connected ground

## **Order Information**

Order Number	Operating Temperature Range	Package	Marking Information	MSL	Transport Media, Quantity
TPAFE411-QFOR	-40 to 125°C	QFN4X4-20	FE411	3	3000

# **Specifications**

# **Absolute Maximum Ratings\***

Parameters	Value	Unit
Power Supply, V <sub>DD</sub> to GND	6.0	V
Input Voltage	GND - 0.3 to V <sub>DD</sub> + 0.3	V
Maximum Junction Temperature, T <sub>J</sub>	150	°C
Operating Temperature Range, T <sub>A</sub>	-40 to 125	°C
Storage Temperature Range, T <sub>STG</sub>	-65 to 150	°C
Lead Temperature (Soldering 10 sec), TL	300	°C

<sup>\*</sup> **Note:** Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

## **ESD, Electrostatic Discharge Protection**

Symbol	Parameter	Condition	Minimum Level	Unit
НВМ	Human Body Model ESD	ANSI/ESDA/JEDEC JS-001	5	kV
CDM	Charged Device Model ESD	ANSI/ESDA/JEDEC JS-002	1.5	kV

## **Thermal Information**

Package Type	$\theta_{JA}$	θ <sub>JC</sub>	Unit
QFN4X4-20	34	35	°C/W



## **Electrical Characteristics**

All test condition is VDD =  $\pm 2.7$  V to  $\pm 5.5$  V,  $T_A = \pm 25$ °C, zero output current IDRV, unless otherwise noted.

		Min	Тур	Max	Unit
Excitation					
	GAIN [00, 01, 10]	0.7	0.8	0.95	V
HALL sensor excitation voltage	GAIN [11]	0.6	0.74	0.95	V
HALL sensor excitation current	T <sub>A</sub> = -40°C to 125°C			10	mA
Excitation switching frequency			1		MHz
	GAIN [00], f <sub>zero</sub> = 3.8kHz		250		V/V
Front-end open-loop flat band gain	GAIN [01], f <sub>zero</sub> = 7.2kHz		250		V/V
	GAIN [10], f <sub>zero</sub> = 3.8kHz		1000		V/V
Front-end open-loop gain	GAIN [00, 01, 10, 11]	91	120		dB
Front and involved to a ffort	No HALL sensor, GAIN [00, 01, 10]		20	100	uV
Front end input voltage offset	GAIN [11]		5	12	mV
Front and involve the man off of hills	No HALL sensor, GAIN [00, 01, 10]		0.2		μV/°C
Front end input voltage offset drift	GAIN [11]		5		μV/°C
Gain bandwidth product	GAIN [11]		14		MHz
Common mode rejection ration	GAIN [11]		300		uV/V
Error comparator threshold			50		mV
	Excitation switching frequency  Front-end open-loop flat band gain  Front-end open-loop gain  Front end input voltage offset  Front end input voltage offset drift  Gain bandwidth product  Common mode rejection ration	$ \begin{array}{c} \text{HALL sensor excitation voltage} \\ \hline & \text{GAIN [11]} \\ \hline \\ \text{HALL sensor excitation current} \\ \hline \\ \text{Excitation switching frequency} \\ \hline \\ \text{Excitation switching frequency} \\ \hline \\ \text{Excitation switching frequency} \\ \hline \\ \text{GAIN [00], f}_{zero} = 3.8 \text{kHz} \\ \hline \\ \text{GAIN [10], f}_{zero} = 7.2 \text{kHz} \\ \hline \\ \text{GAIN [10], f}_{zero} = 3.8 \text{kHz} \\ \hline \\ \text{Front-end open-loop gain} \\ \hline \\ \text{Front end input voltage offset} \\ \hline \\ \text{GAIN [00, 01, 10, 11]} \\ \hline \\ \text{Front end input voltage offset} \\ \hline \\ \text{GAIN [11]} \\ \hline \\ \text{Front end input voltage offset drift} \\ \hline \\ \text{GAIN [11]} \\ \hline \\ \text{Gain bandwidth product} \\ \hline \\ \text{GAIN [11]} \\ \hline \\ \text{Common mode rejection ration} \\ \hline \end{array} $	HALL sensor excitation voltageGAIN [11] $0.6$ HALL sensor excitation current $T_A = -40^{\circ}\text{C}$ to $125^{\circ}\text{C}$ Excitation switching frequencyGAIN [00], $f_{zero} = 3.8\text{kHz}$ Front-end open-loop flat band gainGAIN [01], $f_{zero} = 7.2\text{kHz}$ Front-end open-loop gainGAIN [00, 01, 10, 11]91Front end input voltage offsetNo HALL sensor, GAIN [00, 01, 10]GAIN [11]No HALL sensor, GAIN [00, 01, 10]Front end input voltage offset driftNo HALL sensor, GAIN [00, 01, 10]GAIN [11]GAIN [11]Gain bandwidth productGAIN [11]Common mode rejection rationGAIN [11]	$ \begin{array}{c ccccccccccccccccccccccccccccccccccc$	



# **Electrical Characteristics (Continued)**

All test condition is VDD = +2.7 V to +5.5 V, T<sub>A</sub> = +25°C, zero output current IDRV, unless otherwise noted.

Symbol	Parameter	Conditions	Min	Тур	Max	Unit
Current Sens	ing Amplifier					
Vos	Input voltage offset, RTO	VIN+=VIN-=V <sub>REFIN</sub>		0.01	0.2	mV
V TO	land the second of the DTO	V <sub>IN</sub> =0.5V,1V or 2V,		0.4		\//00
VosTC	Input voltage offset drift, RTO	T <sub>A</sub> = -40°C to 125°C			μν	μV/°C
CMRR	vs common mode voltage, RTO	$V_{CM} = -1V$ to $V_{DD} + 1V$ , $V_{REF} = V_{DD} / 2$ ,		50	500	μV/V
PSRR	vs Power supply, RTO	V <sub>CM</sub> = V <sub>REFIN</sub>		4	50	μV/V
V <sub>CM</sub>	Common mode input range		-1		V <sub>DD</sub> + 1	V
	Differential impedance			12		kΩ
	Common impedance			22		kΩ
	External reference input impedance		36	44	52	kΩ
G	Gain, V <sub>OUT</sub> / V <sub>IN_DIFF</sub>			4		V/V
GE	Gain error			0.02	0.3	%
	Gain error drift			1		ppm//°C
V <sub>OL</sub>	Voltage output swing to ground	I = 2.5mA, comparator trip level		50	150	mV
V <sub>OH</sub>	Voltage output swing to power supply	I = -2.5mA, comparator trip level	V <sub>DD</sub> -150	V <sub>DD</sub> -50		mV
	Short circuit current	V <sub>OUT</sub> connected to GND		-18		mA
I <sub>SC</sub>	Snort circuit current	V <sub>OUT</sub> connected to power supply		20		mA
	Signal overrange indicator delay	V <sub>IN</sub> = 1V step		3		μs
BW <sub>-3dB</sub>	Bandwidth			2		MHz
SR	Slew rate			11		V/µs
	Setting time large signal	ΔV = 2V to 1%		4		μs
	Setting time	ΔV = 0.4V to 0.01%		15		μs
e <sub>n</sub>	Output voltage noise, RTO	f = 1kHz		150		nV/√Hz
Compensatio	n Coil Driver					
	Dook surrent	$V_{DD} = 5V$ , $T_A = -40$ °C to 125°C, $V_{IDRV1}$ - $V_{IDRV2} = 4.2V_{pp}$	300	350		mA
	Peak current	$V_{DD} = 3.3V$ , $T_A = -40$ °C to 125°C, $V_{IDRV1} - V_{IDRV2} = 2.5V_{pp}$	150	200		mA
	Malla na andron	$V_{DD} = 5V$ , $R_{LOAD} = 14\Omega$	4.2			V <sub>pp</sub>
	Voltage swing	$V_{DD}$ = 3.3V, $R_{LOAD}$ = 14 $\Omega$	2.5			$V_{pp}$
	Output common mode			V <sub>DD</sub> / 2		V
			1	1	1	



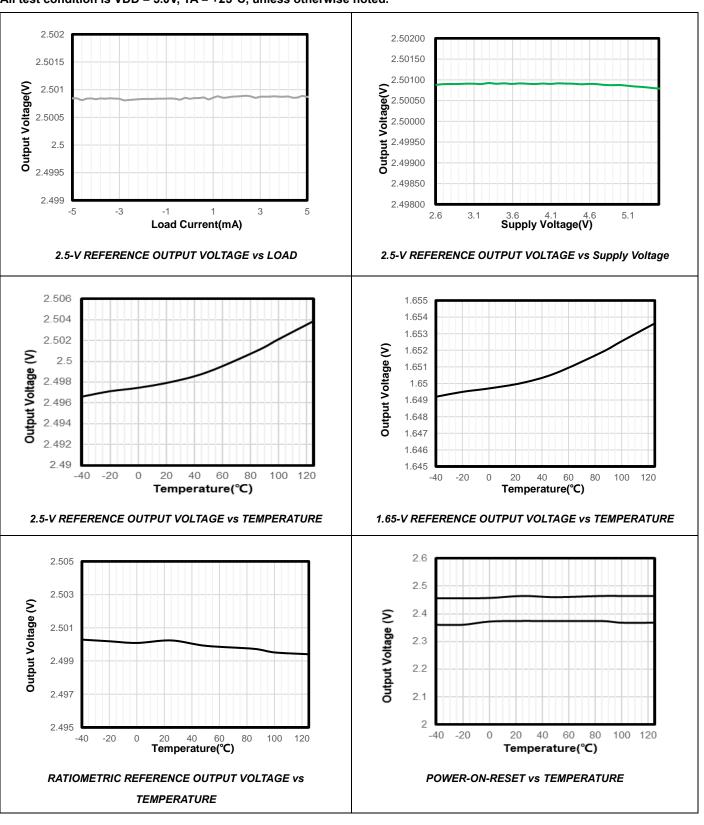
# **Electrical Characteristics (Continued)**

All test condition is VDD = +2.7 V to +5.5 V, T<sub>A</sub> = +25°C, zero output current IDRV, unless otherwise noted.

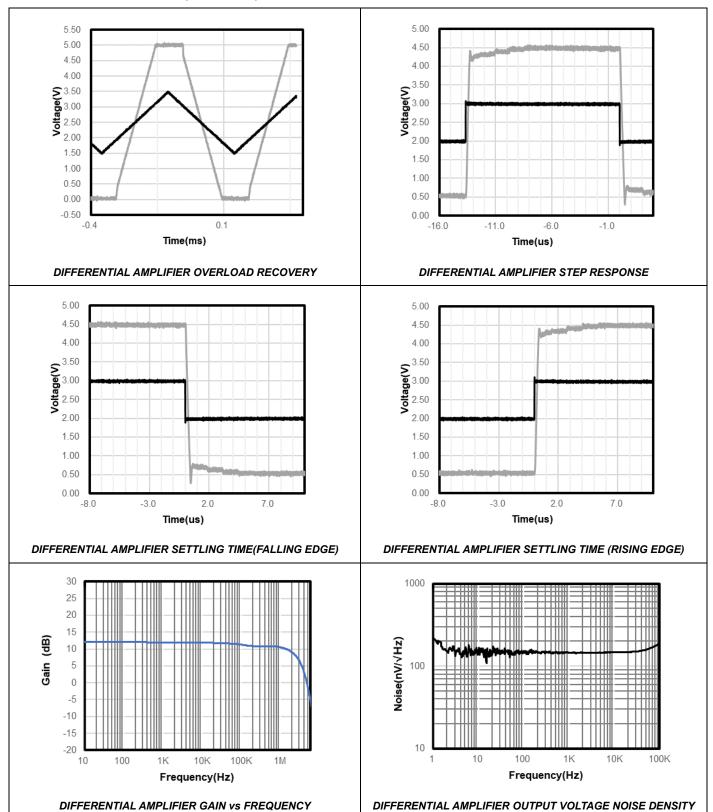
Parameter	Conditions	Min	Тур	Max	Unit
ence					ı
	REFSEL [00]	2.495	2.500	2.505	V
Reference voltage	REFSEL [10]	1.648	1.651	1.654	V
	REFSEL [11]	49.6	50	50.4	% V <sub>DD</sub>
Reference voltage drift	REFSEL [00, 10]		5	50	ppm//°C
Voltage divider gain error drift	REFSEL [11]		5		ppm//°C
Power supply rejection ration	REFSEL [00, 10]		15	200	μV/V
Load regulation	Load to ground or VDD		0.15	0.35	mV/mA
Short circuit current	REFOUT connected to GND		-18		mA
Short circuit current	REFOUT connected to VDD		20		mA
Output					
High level input voltage		0.7 x V <sub>DD</sub>		V <sub>DD</sub> + 0.3	V
Low level input voltage		-0.3		0.3 x V <sub>DD</sub>	V
Low level output voltage	4mA sink		0.3		V
Power supply		2.7		5.5	V
Quiescent current	I <sub>DRV</sub> = 0mA		5.5	8	mA
Power on reset threshold			2.4		V
Specified range		-40		125	°C
Operating range		-50		150	°C
	Reference voltage  Reference voltage drift  Voltage divider gain error drift  Power supply rejection ration  Load regulation  Short circuit current  Output  High level input voltage  Low level output voltage  Low level output voltage  Power supply  Quiescent current  Power on reset threshold  Specified range	Reference voltage  Reference voltage  Reference voltage drift  Reference voltage  Load to ground or VDD  Reference voltage drift  Reference voltag	Reference voltage	Reference voltage	REFSEL [00] 2.495 2.500 2.505 REFSEL [10] 1.648 1.651 1.654 REFSEL [11] 49.6 50 50.4  Reference voltage drift REFSEL [00, 10] 5 50  Voltage divider gain error drift REFSEL [11] 5 5 50  Voltage divider gain error drift REFSEL [11] 5 5 50  Load regulation REFSEL [00, 10] 15 200  Load regulation Load to ground or VDD 15 0.35  Short circuit current REFOUT connected to GND 16 18 18 18 19 19 19 19 19 19 19 19 19 19 19 19 19



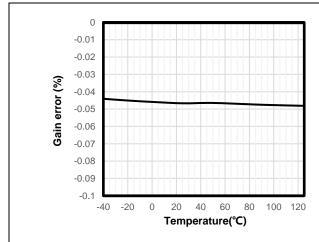
## **Typical Performance Characteristics**



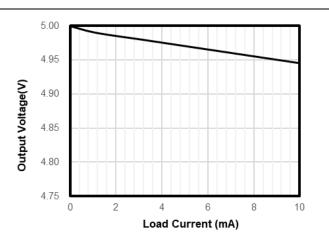
## **Typical Performance Characteristics (Continued)**



## **Typical Performance Characteristics (Continued)**

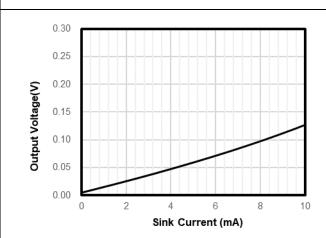


DIFFERENTIAL AMPLIFIER GAIN ERROR vs TEMPERATURE



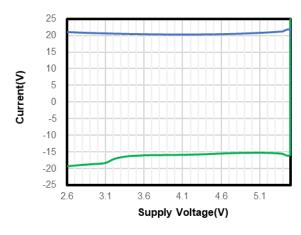
DIFFERENTIAL AMPLIFIER OUTPUT VOLTAGE vs OUTPUT

CURRENT (POSITIVE RAIL)

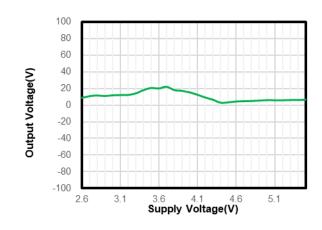


DIFFERENTIAL AMPLIFIER OUTPUT VOLTAGE vs OUTPUT

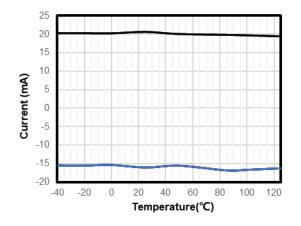
CURRENT (NEGATIVE RAIL)



DIFFERENTIAL AMPLIFIER SHORT-CIRCUIT CURRENT vs POWER SUPPLY



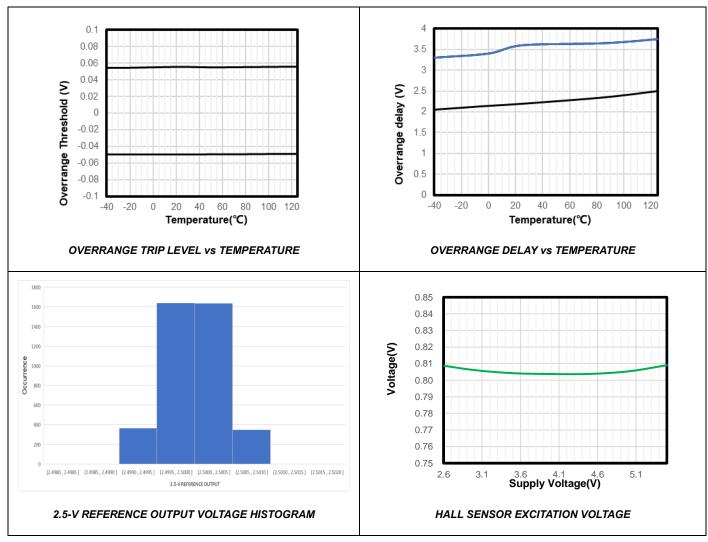
DIFFERENTIAL AMPLIFIER OFFSET VOLTAGE vs POWER SUPPLY



DIFFERENTIAL AMPLIFIER SHORT-CIRCUIT CURRENT vs TEMPERATURE



## **Typical Performance Characteristics (Continued)**





## **Application Information**

TPAFE411 is a signal sensor conditional circuit used to connect current sensor and giving the necessary functions for the sensor operation. The power supply of TPAFE411 is from single +2.7V to +5.5V. And this device includes the basic functions such as magnetic field probe (HALL sensor) excitation, signal conditioning, compensation-coil driver amplification, error condition detecting, and capability of fixing overload circumstance. A precision differential amplifier is embedded to convert the compensation current into output voltage with a shunt resistor. The device also has a precise voltage reference which supplies voltage to comparator, analog-to-digital converter (ADC). A dynamic error correction module is designed to make sure the device keeps the high dc precision and long-term accuracy over the temperature.

The TPAFE411 contains an internal clock and counter logic for managing power-up, overload detection and recovery, error, and timeout situations. In addition, the fabrication of CMOSS process makes the TPAFE411 highly reliable.

### Shunt sense amplifier

To compensate coil of the differential (H-bridge) driver, a differential sense amplifier was required for the shunt voltage. The differential amplifier is designed with wide bandwidth and high slew rate for fast current sensors. chopping is also designed for minimizing the system offset. For gains of 4 V/V, R2/R1=4.

Both inputs of the differential amplifier are tied to the current shunt resistor. This shunt resistor will slight reduce the gain of the amplification circuit and common-mode rejection (CMR). So, a dummy shunt resistor was added in series with the REFin pin to reestablish the matching of both input impedance for the gain of amplification.

Generally, the gain error contributed by the resistance of R-shunt is negligible, but the matching for both resistor divider ratios should be at least higher than 1/3000 for 70dB common-mode rejection.

The output of amplifier drives the input of a Sar-type ADC with an RC low-pass filter. This filter is required to filter out high-frequency component from the amplifier output. For Rf and Cf values, optimum values could be obtained by experiments.

The REFIN pin is the reference node for the output signal (VOUT). The zero reference voltage could be achieved by connecting REFIN to the reference output (REFOUT). The common reference for ADC and TPAFE411 should be used for avoiding the mismatch errors between two reference sources.

#### Overage comparator

If there is an overload current flowing the shunt resistor, the OR pin will be pulled low to indicate an overvoltage condition for the differential amplifier. The output pin will hold 3us before flip-flop in case the noise triggers the flag. The OR pin will return to high as no overload current exists. This error flag will warn system to shut down the circuit. And the shunt resistor value defines the working condition of current sets the ratio between the nominal signal and triggered level of the overload flag. This trip current is calculated with following examples:

The output voltage swing is approximately  $\pm 2.45$  V (load and supply voltage-dependent) at a 5-V supply. Divide by the gain of 4 V/V, an input swing is  $\pm 0.6125$  V, and the clipping current is IMAX = 0.6125 V / RSHUNT.

The over range condition is measured immediately when amplifier approaches the rail and exceeds the linear operating range. Therefore, the error flag of the over range comparator level can indicate the fault circumstances such as output shorts, low load, or low-supply conditions. The flag signal will keep active if the output can't drive the voltage higher.

#### Voltage reference

The precise voltage reference supplies low drift voltage and is used to bias the internal circuit, it is also tied to the REFOUT pin. The circuit works as the reference point of the output signal to allow a bipolar signal around it. The output has the internal buffer for low impedance and allows maximum+-5mA current to be sank or sourced.

Capacitive loads can be connected directly but may have the ringing for fast load transients. To achieve the better transient response, a small series resistor can be placed in series.



### Reference output voltage selection

As shown below, the 5V and 3.3V power supplies are usually selected for the TPAFE411. The sensor output must be set at 2.5V and 1.65V, respectively. The internal reference gives the low drift and precise reference voltage.

MODE	REFSEL1	REFSEL2	DESCRIPTION
REF = 2.5 V	0	0	Used with sensor module supply of 5 V
REF = 1.65 V	1	0	Used with sensor module supply of 3.3 V
Ratio-metric output	1	1	Provides output centered on V <sub>s</sub> / 2

For the ratio-metric mode, the reference is bypassed, and the power supply is divided by two. The internal resistor divider gives a very good temperature coefficient which is less than 10ppm/. And the sensor output is around VS/2 in this case.

### Power-On startup and brownout

Power-on is activated when the power supply goes above 2.4V. At this point, digital logic begins to work and waits for 100us for power supply to settle. During this time, ICOMP1 and ICOMP2 outputs are pulled down to low in order that there is no undesired signal driving the compensation coil. The ERROR pin will hold low for 100us when error happens in case that false error triggers the output. VOUT will be only valid for 100us after power-on reset.

The brownout voltage level of TPAFE411 is around 2.4V. Bypass capacitors and stable power supply are required for driving the heavy current by TPAFE411. The supply voltage drop that lasts longer than 25us will activate the power-on reset. If the voltage drops below 1.8V, the power-on reset will be also triggered. When the power supply returns to 2.4V, the device will restart the startup process as described above.

#### **Error condition**

If there is a signal clipping in the differential amplifier, the over range flag will be activated. It will also trigger the system error flag. This error points out that the output voltage does not represent the primary current. The error flag is activated when the power fails and browns out or Hall sensor is not within its normal operating range.

Both error and over range flags are based on the open-drain circuit, an external pull-up resistor is required.

The conditions listed below will activate the error flag:

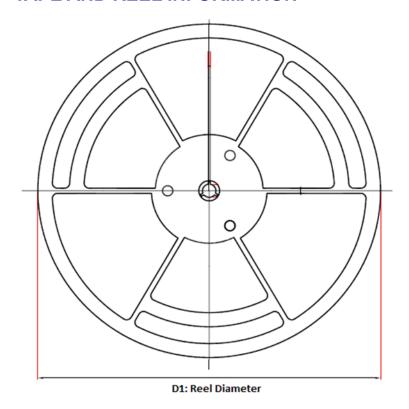
- 1. The Hall sensor offset was bigger than 50mV.
- 2. Any terminal of Hall sensor is disconnected.

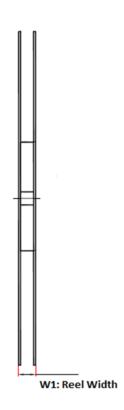
#### **Protection recommendations**

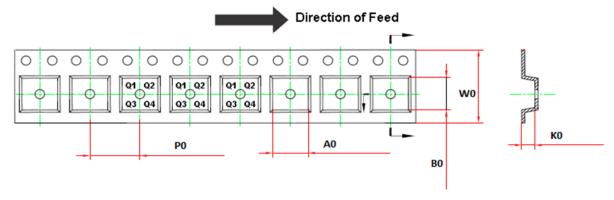
The inputs IAIN1 and IAIN2 need external protection to limit the voltage swing below 6 V of the voltage supply. ICOMP1 and ICOMP2 can afford high-current pulse due to internal clamp circuit. Schottky diodes should be connected to the supply rail when large currents are expected.



# TAPE AND REEL INFORMATION





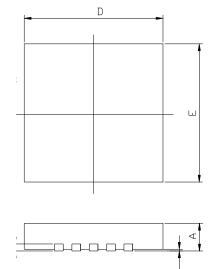


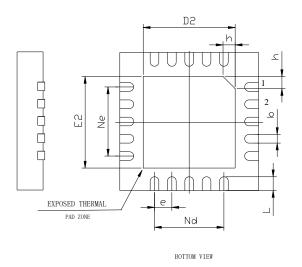
Order Number	Package	D1	W1	A0	В0	K0	P0	W0	Pin1
		(mm)	(mm)	(mm)	(mm)	(mm)	(mm)	(mm)	Quadrant
TPAFE411-QFOR	QFN4X4-20	330.0	17.6	4.3	4.3	1.1	8	12.0	Q2



# **Package Outline Dimensions**

## QFN4X4-20





SYMBOL	MILLIMETER		
	MIN	NDM	MAX
Α	0.70	0.75	0.80
A1		0.02	0.05
b	0.18	0.25	0.30
C	0.18	0.20	0.25
D	3.90	4.00	4.10
D2	2.55	2.65	2.75
е	0.50BSC		
Ne	2.00BSC		
Nd	2.00BSC		
E	3.90	4.00	4.10
ES.	2.55	2.65	2.75
Ĺ	0.35	0.40	0.45
h	0.30	0.35	0.40
L/F 载体尺寸 (mil)	114X114		





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